

## Internal impedance of a

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### Conducting medium:

The intrinsic impedance or characteristic impedance of a non-conducting medium is found to be simply square root of the ratio of parameters of medium namely  $\mu$  &  $\epsilon$

For free space it is equal to  $377\Omega$

i.e. The impedance  $Z$  or  $\eta = \frac{E}{H} = \sqrt{\frac{\mu}{\epsilon}}$  ( $\Omega$ )

for free space  $\mu = \mu_0$  &  $\epsilon = \epsilon_0$  (units)

$$Z = \eta = \sqrt{\frac{\mu_0}{\epsilon_0}} = 377\Omega \quad \text{As } \frac{E}{H} = \frac{V/m}{A/m} = \frac{V}{A} = (\Omega)$$

So for free space  $377\Omega$  is purely resistive.

If a medium is a conducting one that is having finite conductivity, the impedance of the medium is no more pure resistive but becomes a complex factor.

An expression for the impedance can be derived as follows:

Considering the Maxwell's equations for

$$\vec{E} \text{ \& \ } \vec{H}, \quad \nabla \times \vec{E} = -j\omega\mu\vec{H} \quad (1)$$

$$\nabla \times \vec{H} = (\sigma\vec{E} + j\omega\epsilon\vec{E}) \quad (2)$$

Hints

While taking

$$\vec{B} = \vec{B}_0 e^{j\omega t}$$

$$\vec{E} = \vec{E}_0 e^{j\omega t}$$

$$\vec{B} = \mu\vec{H}, \quad \vec{J} = \sigma\vec{E}$$

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For a uniform plane wave propagating along let us say in the  $x$ -direction

The  $yz$ -plane directions of  $\text{Curl of } \vec{E} \text{ \& } \vec{H}$  can be expressed as

$$(\vec{\nabla} \times \vec{E})_{yz} = -\frac{\partial E_z}{\partial x} \hat{j} + \frac{\partial E_y}{\partial x} \hat{k} \quad (3)$$

Note we take in the  $\vec{\nabla} \times \vec{E}$  the terms of  $E_y$  which vary with " $x$ " while all other  $E$  terms considered constants & their partial derivatives are considered zero.

$$(\vec{\nabla} \times \vec{H})_{yz} = -\frac{\partial H_z}{\partial x} \hat{j} + \frac{\partial H_y}{\partial x} \hat{k} \quad (4)$$

Substituting equations (3) & (4) in equations (1) & (2) respectively

$$-\frac{\partial E_z}{\partial x} \hat{j} + \frac{\partial E_y}{\partial x} \hat{k} = j\omega\mu (H_y \hat{j} + H_z \hat{k}) \quad (5)$$

Where  $\vec{E}_{yz} = E_y \hat{j} + E_z \hat{k}$ , &  $\vec{H}_{yz} = H_y \hat{j} + H_z \hat{k}$

$$-\frac{\partial H_z}{\partial x} \hat{j} + \frac{\partial H_y}{\partial x} \hat{k} = (\sigma + j\omega\epsilon) (E_y \hat{j} + E_z \hat{k}) \quad (6)$$

Equating the terms in the equations (5) & (6),

& re-arranging we get

$$-\frac{\partial H_z}{\partial x} = (\sigma + j\omega\epsilon) E_y \quad (7)$$

$$\frac{\partial E_z}{\partial x} = j\omega\mu H_y \quad (8)$$

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$$\frac{\partial H_y}{\partial x} = (\sigma + j\omega\epsilon) E_z - j, \quad (3/6)$$

$$\frac{\partial E_y}{\partial x} = -j\omega\mu H_z \quad (10)$$

Dividing equation (7) by (10) we have

$$\frac{\partial E_y}{\partial H_z} = \frac{j\omega\mu}{\sigma + j\omega\epsilon} \left( \frac{H_z}{E_y} \right) \quad (11)$$

Re-arrange eq (11)

$$E_y \partial E_y = \frac{j\omega\mu}{\sigma + j\omega\epsilon} H_z \partial H_z \quad (12)$$

Integrating the both sides of eq (12) & ignoring constant

$$\int E_y \partial E_y = \frac{j\omega\mu}{\sigma + j\omega\epsilon} \int H_z \partial H_z$$

$$\frac{(E_y)^2}{2} = \frac{j\omega\mu}{\sigma + j\omega\epsilon} \frac{(H_z)^2}{2}$$

$$\left( \frac{E_y}{H_z} \right)^2 = \frac{j\omega\mu}{\sigma + j\omega\epsilon}$$

$$\frac{E_y}{H_z} = \sqrt{\frac{j\omega\mu}{\sigma + j\omega\epsilon}} \quad (13)$$

Hence the ratio of  $\vec{E}$  to  $\vec{H}$  is written as

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$$Z = \eta = \frac{E}{H} = \sqrt{\frac{j\omega\mu}{\sigma + j\omega\epsilon}} \quad (14) \quad \boxed{(4/6)}$$

which is the impedance of the conducting medium with finite conductivity.

for good conductors  $\sigma \gg \omega\epsilon$  the

expression (14) reduces to

$$\eta = \sqrt{\frac{j\omega\mu}{\sigma}} \quad (15)$$

$$\eta = \sqrt{j} \left( \sqrt{\frac{\omega\mu}{\sigma}} \right) \quad \sqrt{j} = \left( \frac{1+j}{\sqrt{2}} \right)$$

$$\eta = \left( \frac{1+j}{\sqrt{2}} \right) \left( \sqrt{\frac{\omega\mu}{\sigma}} \right)$$

$$\eta = \sqrt{\frac{\omega\mu}{2\sigma}} + j \sqrt{\frac{\omega\mu}{2\sigma}}$$

$$\eta = \sqrt{\frac{\omega\mu}{\sigma}} \angle 45^\circ \quad (16)$$

From the equation (16) it follows that the magnitude of intrinsic impedance of a conductor is given by  $\eta = \sqrt{\frac{\omega\mu}{\sigma}}$  with the phase difference of  $45^\circ$ .

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At a frequency of 3 GHz the magnitude of intrinsic impedance for copper is found to be  $0.020 \Omega$

$$(\eta)_{Cu} = 0.020 \Omega$$

This shows that for a conducting medium such as copper the ratio  $E$  to  $H$  is much less than for free space.

for a perfect conductor  $\sigma$  is infinite & "  $\eta$  " i.e. intrinsic impedance tends to zero hence  $E$  vanishes  $\vec{E} = 0$

The small value of "  $\eta$  " suggests that a conducting medium like copper behaves as a short circuit to EM-fields

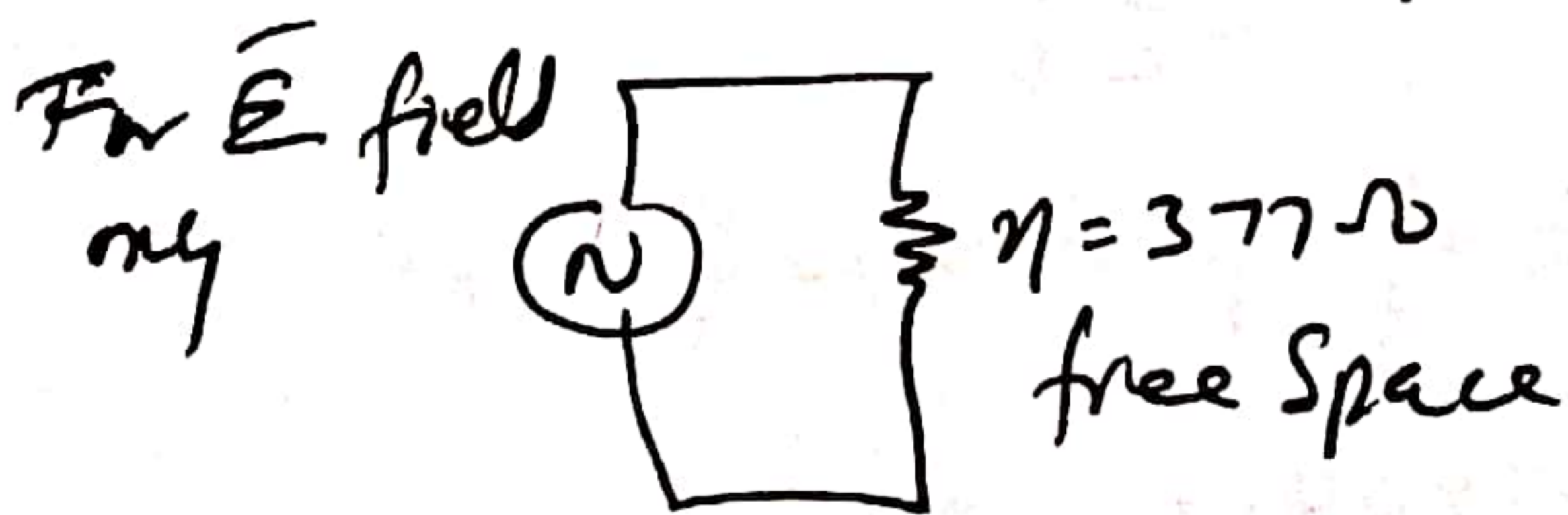


Fig-a Fields propagate in free space

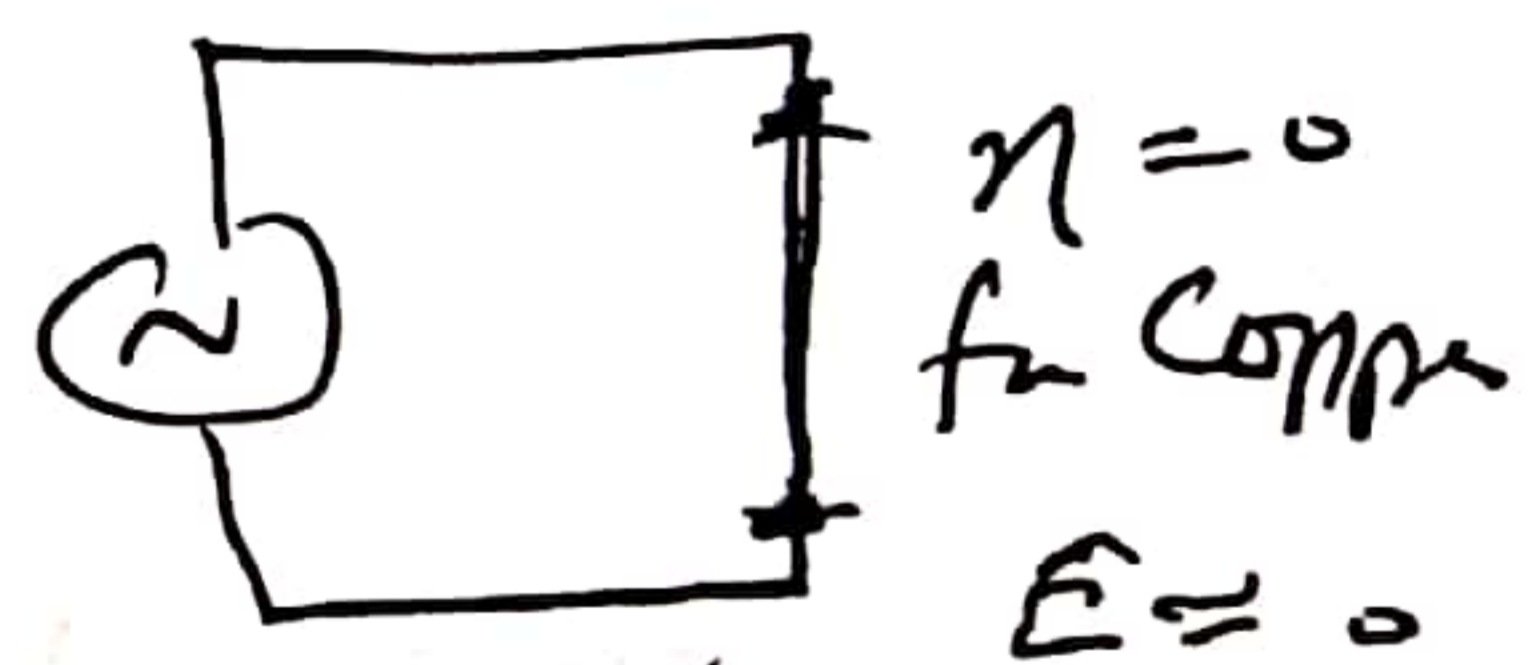


Fig-b Fields vanish at surface of copper medium

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SNo	Element	Conductivity $\sigma \times 10^7$ (S/m)	$\mu \times 10^{-7}$ $\mu$ (H/m)	Skin depth $\delta$ (m)
1	Silver	6.17	$4\pi$	$\frac{0.0642}{\sqrt{f}}$
2	Copper	5.80	$4\pi$	$\frac{0.0660}{\sqrt{f}}$
3	Al	3.72	$4\pi$	$\frac{0.0826}{\sqrt{f}}$
4	Brass	1.57	$4\pi$	$\frac{0.127}{\sqrt{f}}$

Take an example of Copper  $\text{Cu}$   
 At frequency,  $f = 50 \text{ Hz}$  for Skin depths,

$$\delta = \frac{0.0642}{\sqrt{50}} = 0.0093 \text{ m} = 9 \text{ mm}$$

At frequency,  $f = 3 \text{ GHz}$

$$\delta = \frac{0.0642}{\sqrt{3 \times 10^9}} = \underline{1.2 \mu\text{m}}$$

The impedance at various frequencies

$$\eta = \sqrt{\frac{\omega \mu}{\sigma}} = \sqrt{\frac{2\pi f \mu}{\sigma}}$$

For 50 Hz

$$\eta = 2.6 \times 10^{-6} \Omega$$

For 3 GHz

$$\eta = 0.02 \Omega$$

Solve for  $\delta$  &  $\eta$  for other elements in the table